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E-BROIDERY: AN INFRASTRUCTURE FOR WASHABLE COMPUTING

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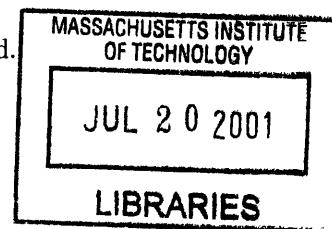
E. Rehmi Post

B.S., Physics
University of Massachusetts, Amherst
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Submitted to the Program in Media Arts and Sciences, School of Architecture and Planning, in partial fulfillment of the requirements for the degree of Master of Science in Media Arts and Sciences at the Massachusetts Institute of Technology

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Author

E. Rehmi Post
Department of Media Arts and Sciences
October 9, 1998

Certified by

Neil Gershenfeld
Associate Professor of Media Arts and Sciences
MIT Media Lab

Accepted by

Stephen A. Benton
Professor of Media Arts and Sciences
Chair, Departmental Committee on Graduate Students
Program in Media Arts and Sciences

ROTCH

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Finally, I dedicate this work to Jenny Gutbezahl, whose unfailing love and caustic wit have somehow conspired to keep me true.

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Abstract

Wash-and-wear multilayer electronic circuitry can be constructed on fabric substrates, using conductive textiles and suitably packaged components. Fabrics are perhaps the first composite materials engineered by humanity; their evolution led to the development of the Jacquard loom, which itself led to the development of the modern computer. The development of fabric circuitry is a compelling closure of the cycle that points to a new class of textiles which interact with their users and their environments, while retaining the properties that made them the first ubiquitous “smart material”. Fabrics are in several respects superior to existing flexible substrates in terms of their durability, conformability, and breathability. The present work adopts a modular approach to circuit fabrication, from which follow circuit design techniques and component packages optimized for use in fabric-based circuitry, flexible all-fabric interconnects, and multilayer circuits. While maintaining close compatibility with existing components, tools, and techniques, the present work demonstrates all steps of a process to create multilayer printed circuits on fabric substrates using conductive textiles.

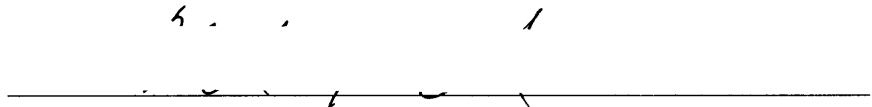
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E-BROIDERY: AN INFRASTRUCTURE FOR WASHABLE COMPUTING

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Adviser

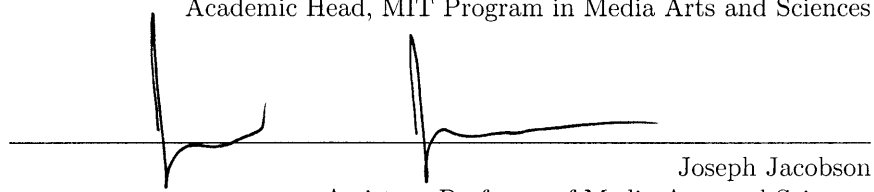


Neil Gershenfeld
Associate Professor of Media Arts and Sciences
Director, Physics and Media Group
Thesis Supervisor

Reader

Alex Pentland
Professor of Media Arts and Sciences
Director, Vision and Modelling Group
Academic Head, MIT Program in Media Arts and Sciences

Reader



Joseph Jacobson
Assistant Professor of Media Arts and Sciences
Director, Micromedia group

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Chapter 1

Introduction

One of the few constants in modern society is that everybody wears clothing. Clothing serves many functions: it provides us with a sense of the wearer's identity; it affords protection from the environment; and it gives the wearer pockets or other convenient places to store small objects.

All of these functions are also served by wearable computing [SMR⁺96] in a digital analog. If the user inhabits an information-rich environment a wearable computer can provide others with chosen details of the user's identity and preferences. It can protect the user by mediating his/her experience and refining a torrent of raw data into a salient stream of information, linked to the immediate physical and informational context. And it provides a place to store informational objects the user might wish to use later.

Of course, wearable computing has a number of precedents. Since the beginning of the digital era computers have followed a trend of becoming smaller, faster, and more personalized. In the early 1980's the personal computer revolution began to put computers on every desk, while cellular telephones and pagers changed personal communications. In the late 1990's, personal, miniaturized digital systems have become increasingly important in many people's everyday lives. With the privatization of the Internet and the birth of the World Wide Web, "being connected" has emerged as a popular concern.

But the electronic gear that makes this information revolution possible still takes an awkward form. It's not truly wearable except in the sense that it will fit in your pocket or strap onto your body. What we need is a way to integrate technology directly into clothing.

The goal of this work is to make electronic circuitry disappear into clothing, by replacing circuit boards with fabric circuitry and by using component packages designed for use in fabric circuits. Several of the reasons to do this are outlined below, all stemming from the desire to make e-garments with integrity. An e-garment's integrity depends on its ability to look, feel, wash, and wear as well as

ordinary clothing does.

Circuit boards do not look or feel good when they become part of your clothing, nor do they wash well. We might address this by making the circuit board smaller and removable, by redesigning it to use the same components in smaller packages and by using conductive fasteners to connect boards and garments (so that boards can be removed when a garment goes into the wash.) The trouble is, this approach doesn't scale well: the whole circuit board still has to be assembled and tested, and somehow connected to the fabric. Worst of all, there are still several layers of packaging between the silicon and the silk.

All of this points to a better way of doing things. Rather than choosing packages to increase component density on a circuit board, one might instead choose packages to do the same thing in a fabric circuit. One might also choose to package components so they can be washed without compromising their electrical connection to the garment and to other components. Finally, the interface between components and circuitry must withstand the flexing and stretching that clothing is prone to experience.

1.1 How to make washable circuitry

Commonly, printed circuit boards (PCBs) have two or more layers of printed conductive traces which interconnect components soldered onto exposed portions of the outermost layers. This method of circuit fabrication is pervasive, so it would be good for any technique we develop be compatible with existing printed circuit layout tools. Consequently, we adopt a modular approach in which components and printed circuit strata are produced separately and then joined in standardized ways to produce circuits of arbitrary complexity.

The first part of this work is to identify good materials for use in fabric circuitry. This involves a tradeoff between electrical mechanical properties: we want to use a thread that is reasonably conductive but that can be sewn at high speed in commercial embroidery processes. Such an approach builds on prior art in textile manufacturing to develop a composite yarn that balances these properties and is also washable. For the purposes of this work, we consider dry-cleaning as the target washing process, and exposure to water, alcohols, and sweat as an everyday occurrence.

The second part of this work shows how to design packages optimized for use in fabric circuitry. The focus of this effort is the design of the "Plastic Threaded Chip Carrier" (PTCC), a package that bears a resemblance to the Plastic Leadless Chip Carrier (PLCC). The PLCC is a well-known surface-mount IC package style, designed to be soldered onto printed copper circuit traces on a rigid or semi-rigid circuit board. Conversely, the PTCC is designed to be stitched or woven into a fabric circuit, and for this purpose has long, flexible conducting leads.

The third part of this work is to develop a way to sew multi-layer circuitry into fabric. Even if we can sew a circuit, its complexity will be limited if it is restricted to two dimensions. This part of the present work builds on existing techniques on embroidery that achieve particular effects of texture and appearance.

The goal is to demonstrate the feasibility of wash-and-wear articles that integrally include digital circuitry, to provide data on the lifetime of such circuitry, and to describe the particulars of the relevant fabrication methods. Ultimately we would like to demonstrate that textile-based circuitry compares favorably to ordinary printed circuit boards in terms of its complexity and reliability. It must also be more durable than printed circuit boards are when exposed to the mechanical, thermal, and other environmental stresses that clothing and wearable items normally experience.

Beyond the applications to clothing, this work will make possible the construction of circuits with a high degree of flexibility including user interfaces that conform to irregularly shaped surfaces.

1.2 A wash-and-wear digital infrastructure

Imagine wearing your computer all the time. Your interactions with it would not require a rigid, structured form but would instead incorporate your motions, your gestures, and aspects of your affect and environment as a context for the interaction [SMR⁺96].

Giving the computer this detailed level of awareness of its user is akin to giving the computer a "skin" in the real world. Think of how we interact through our interfacial layer with the world around us. In one sense, we are proposing to merge the computer's skin with our clothing. This idea has a precedent in the AI community, particularly in the work of Rodney Brooks, who suggests that behavior that seems to be "intelligent" evolves as a response to sensory presence in a physical environment.

Merging computing and clothing is in some ways also a departure from the traditional view of the cyborg: a half-human, half-cybernetic organism where biological and digital systems combine in synergy. These "borgs" often have an imposing presence, if only because they appear to live within a carapace of impact-resistant plastic, heavy-duty wiring, and obvious bulges of siliconic ganglia. This is really a difference in fashion and appearance, not in motivation. What we propose here is a way to unobtrusively integrate digital systems into clothing.

Portable phones and pagers are two kinds of personal digital systems that have benefited from intense efforts at miniaturization. Due to the economies of scale, the cost of manufacturing these devices has fallen to the point where it is "in the noise" compared to the cost of maintaining the infrastructure necessary for their operation. By increasing the potential user base to anyone who wears clothes, the cost of devices integrated into clothing could be lowered even farther.

One might imagine a wearable computer composed of several parts, each integrated into a separate article of clothing. For example, it has been proposed that shoes are a good place to put computing equipment, because of the space available in the shoe itself as well as the power that potentially can be extracted from the human gait. The shoe is also a good location for a system to transmit and receive data [Zim95] and power [PRPG97] through the body.

1.3 Previous Work in Conductive Textiles

The state of the art in textile circuitry seems to be largely unpublished and proprietary. At the time of this writing, I have not been able to find any prior art relating to the construction of actual circuitry in fabric using off-the-shelf electronic components and ordinary fabric. While a great deal of work has been done in the field of electrically active textile technology, it appears that almost none at all has been done in the construction of active electronic circuits on or in fabric.

An inspirational exception to this was an experiment conducted by Michael Hawley at the Media Lab, which (among other things) sought to impart electronic circuitry into fabric. This path was abandoned because the appropriate materials and techniques were not identified. The development of these methods is informed by both the properties of the materials used and by the capabilities of printed circuit board fabrication technology.

There are, of course, many related threads of ongoing work. A survey of the U.S. Patent literature reveals that most workers are focusing on the development of synthetic conductors compatible with existing textile production processes. For example [GKK89], there are certain conductive polymers which retain a comparatively high conductivity (approx. $1\Omega \cdot m^{-1}$) after being spun into threads which have the weight, strength, and appearance of ordinary nylon thread.

Another knot of work focuses on the production of conductive materials and processes for printing them onto fabric [GGJ92]. Workers in this area cite the possibility of “tagging” apparel by printing tuned LC circuits directly onto the fabric, rather than by attaching separate, removable anti-theft tags. This problem is also addressed by workers who seek to embed magnetostrictive tags directly into clothing but this is not directly related to constructing circuitry directly in clothing.

Still others are working on embedding photonic conductors (optical fibers) into clothing to build architectures for sensing and data distribution. The sensate liner [LEB⁺97] lies in this strand, in which a woven fiber optic matrix is used to sense disruptions of the weave (by measuring optical continuity) and to perform spectrographic analysis of the materials present at intrusions. Such analysis can be used to determine the type and severity of wounds suffered by the wearer of the liner. This line of work is promising because of the variety of fiber optic sensors that can be made, but their largest drawbacks arise from the interface to conventional electronics that are required for signal origination

and processing.

1.4 Case study: The Wearable MIDI Jacket

This wearable system integrates a conductive fabric keypad, a MIDI sequencer/synthesizer, and loudspeakers into a denim jacket. The jacket demonstrates that keyboards can be made in a single layer of fabric using capacitive sensing [Bax97], where an array of embroidered or silk-screened electrodes make up the points of contact. A finger's contact with an electrode can be sensed by measuring the increase in the electrode's total capacitance. It is worth mentioning that this can be done with a single bidirectional digital I/O pin per electrode, and a leakage resistor sewn in highly resistive yarn. Capacitive sensing arrays might also be used to tell how well a piece of clothing fits the wearer, because the signal varies with pressure.

The keypad shown here has been mass-produced using ordinary embroidery techniques and mildly conductive thread. The result is a keypad that is flexible, durable, and responsive to touch. A printed circuit board supports the components necessary to do capacitive sensing and output keypress events as a serial data stream. The circuit board makes contact with the electrodes at the circular pads only at the bottom of the electrode pattern. In a test application, 50 denim jackets were embroidered in this pattern. Some of these jackets are equipped with miniature MIDI synthesizers controlled by the keypad. Although no user studies have been conducted, the responsiveness of the keyboard to touch and timing were found by several users to be excellent.

A view of the component side of the circuit board has been superimposed to show its extent and its connections to the fabric. A flexible circuit board can be substituted for the rigid one used in this implementation. The work which follows describes how to do away with the circuit board entirely.



Figure 1.1: The Wearable MIDI Jacket.

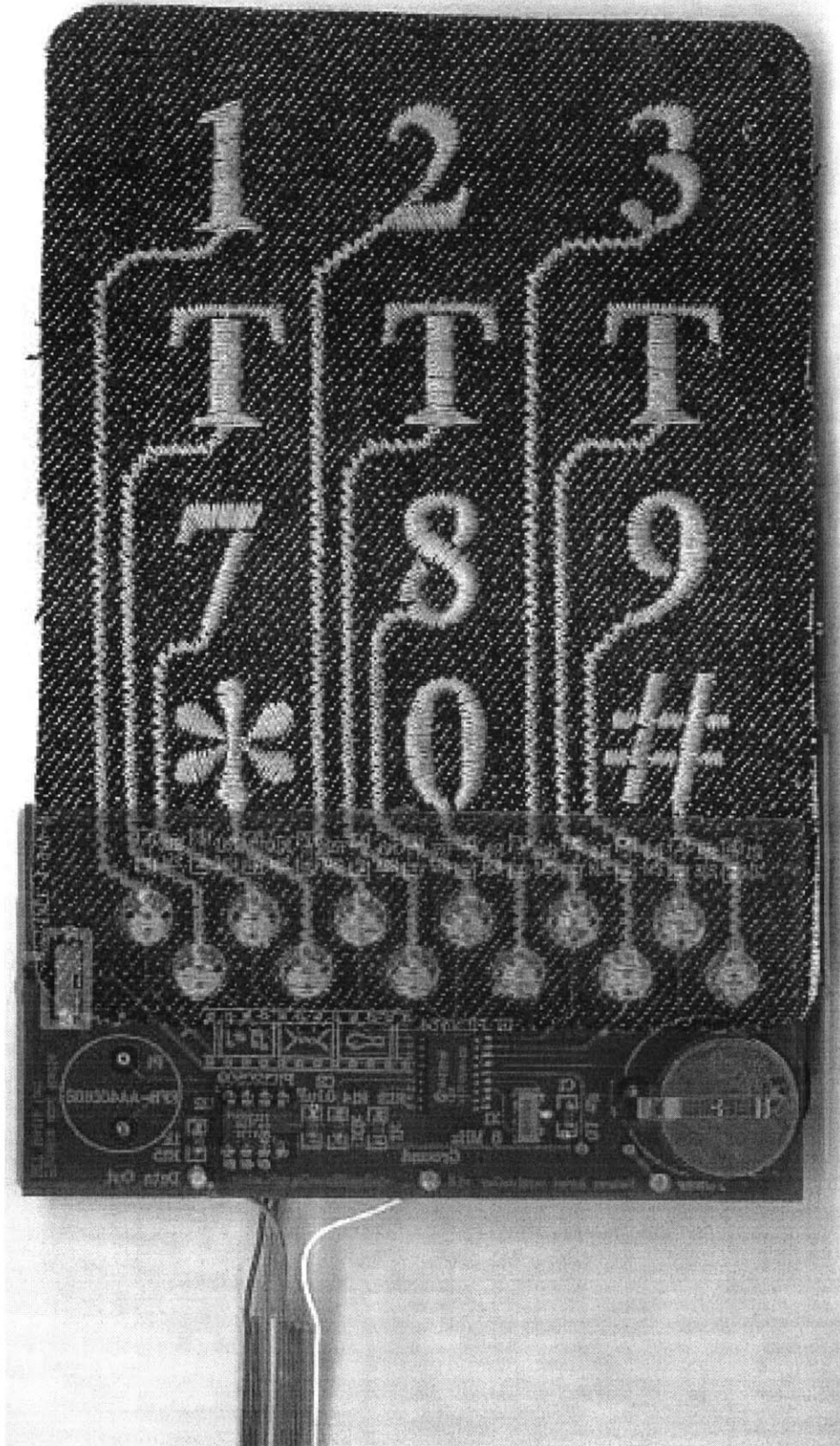


Figure 1.2: Composite image of denim keypad and readout circuit board.

Chapter 2

Context and Implementation of Work

Electrical circuits are typically assembled by soldering active and passive electrical components onto solid boards. The components receive power and exchange signals by means of a network of conductive metal traces patterned in multiple layers on the board.

This approach to circuit fabrication is nearly universal, but it limits the form taken by electronic devices. Rigid boards must be housed within rigid cases, so the notion of electronics being packaged in "boxes" is ubiquitous and alternatives are difficult to imagine.

As the miniaturization of circuits continues, and as the suite of electronically useful materials expands, alternatives to the tradition of rigid housings and circuit boards will become increasingly important. At the time of this writing, most consumer electronics are housed in so-called "biomorphic" packages designed to have smooth curves rather than hard edges.

Current work into user interface design attempts to bring electronic sensing and display circuitry into more intimate contact with users. Interaction with digital objects is likely to become as common as interaction with objects that are not informationally enhanced. In this way, the user is spared the need to deliberately "operate" an external system.

The range of useful tasks amenable to digital mediation is likewise increased. In this vision, environment and location monitoring, information storage, processing and mediation, and short- or long-range digital communication may all be effected without effort by the user or proximity to external electronic devices.

2.1 Abstraction

Most integrated circuits and discrete components are available in standard packages meant to attach to printed circuit boards. These components are usually soldered into place, although conductive adhesives have recently shown promise as alternatives to solder.

To be compatible with current components, we would like to preserve certain properties of printed circuit boards when building circuits in fabric, as well as drawing upon existing knowledge in flexible circuit fabrication.

- *Electrical properties:* The versatility of printed circuits is due in part to their applicability over a broad range of frequencies, from zero to several gigahertz. At low frequencies, the circuit designer is free to consider the printed traces to be nearly ideal conductors, much like the lines connecting components in a circuit diagram. At higher frequencies, however, the traces themselves become circuit elements, as their feature size becomes similar to the length of the signals they carry.
- *Mechanical properties:* Printed circuits often comprise copper bonded to some substrate, such as a fiberglass/epoxy hybrid. The substrate is strong, flexible, and electrically well- characterized. It has good dimensional stability, and damps mechanical vibration. There are more exotic substrates, optimized for flexibility (Kapton, for example) or connection density (multi-chip modules).
- *Component packaging and bonding:* The mechanical properties of a printed circuit also influence the ways in which components can be connected to it. The best way to connect integrated circuits (ICs) is to wirebond them directly to one another or to an interconnecting substrate. This can be an exacting process, so most components are instead mounted in packages of standard shape, size, and pin arrangement. These packages trade off larger connections and more tolerance for error against connection density and parasitic electrical characteristics. Recently, the Ball Grid Array has been promoted as a high-density connection scheme, in which an array of pads on an IC's surface connect directly to a printed circuit through small metal balls and an optional protective layer.

It is alluring to think of woven circuitry, where components are formed by the junctions of individual threads with different material properties. As a modest and practicable first step toward that end, we propose to address the properties of substrates, interconnects, and component packages in the context of textile-based circuit fabrication.

Interconnections must be both mechanically and electrically stable, at the interface between component and textile, and along the run of the interconnecting textile. This requirement affects the

bonding strategy, the component packaging, and the choice of material. We will explore each of these areas in turn, by designing and evaluating

- methods for component bonding,
- packages and lead patterns optimized for use with textiles, and
- composite materials designed with strength, conductivity, and washability in mind.

2.2 Implementation Goals

Rather than confronting the problem of synthesizing conductive polymers, one may instead relax the requirement of highly conductive interconnections, and instead use circuit elements that have low power requirements and high input impedance. This allows one to bootstrap the evolution of fabric circuitry. Such an approach led to the construction of fabric interconnects and sensing arrays in the Wearable MIDI Jacket.

The primary goal is to demonstrate a durable interconnection scheme for components in fabric, replacing printed circuits by stitched circuits. It has already been mentioned that material properties are crucial in this regard. It is also important to consider the patterns in which conductive threads are stitched, because this also strongly determines the electrical properties of the resulting circuitry.

This also points out the possibilities of replacing discrete components such as capacitors, resistors, and inductors by specific combinations of thread and stitch pattern. This can lead to reduced manufacturing cost and higher reliability, but it also is a starting point for the discussion of components formed by combinations of different materials.

Another form of stitched component is the sensor. In the case of the wearable MIDI jacket, stitching forms both an interconnect and an array of capacitive sensors, with the distinction arising from differences in contact area and material properties influenced by the stitch pattern. Sensing can also be done by the thread itself, or by the materials which permeate the thread.

The primary circuit placement method is embroidery with conductive thread, or e-broidery. In this case, conductive thread is stitched in patterns defining conductive regions, to produce circuit traces, component connection pads, or sensing surfaces. So far this work has been limited to a single layer of stitching, but it may be possible to build a repertoire of multi-layer embroidery techniques. The simplest test of this would be to embroider a pair of traces where one crosses over the other without direct electrical connection between the two. A more ambitious idea would be to embroider multilayer sensing structures, such as a normally-open electrical switch, where an intermediate layer separates two conducting stitches unless the top conductor is pressed.

The foremost goal of this inquiry has been to generate the tools and processes whereby a designer may construct e-garments that can withstand daily use and dry cleaning. Rather than seeking to supplant conventional packaging and construction of wearable computers, we start instead at the boundary between users and their computers and describe a way of constructing interfaces that have the familiar look and feel of clothing.

In this context, system integration could mean partitioning a digital system across several items of clothing. Part of this is trading functionality against cost. For example, most people have only a few pairs of shoes, a few more pairs of pants, several shirts, and a few outer garments. So it makes sense to put the most expensive parts of a system in the shoes and outer garments, and the least expensive parts in the shirt and pants.

The main body of this work therefore lies in describing new form factors for components destined for use in wash-and-wear fabric circuit substrates, as well as a means of constructing non-trivial circuits on fabric substrates that retain the appearance of ordinary wearable cloth. Component packages are now optimized to interface with printed circuit boards, so we instead optimize them to interface with textile circuitry. Ultimately, we hope this work will lead to an actual packaging standard used by the integrated circuit industry.

Chapter 3

Fabrication techniques

There are several ways to embed electronic circuitry in fabric, depending on the choice of substrate. Some possibilities explored in the course of this work include

- soldering surface-mount components directly onto metallic organza,
- bonding components to a substrate using conductive adhesives,
- “stapling” components into a conductive stitched circuit (pressure-forming their leads to grip circuit pads), and
- couching a component’s threadframe directly into a circuit (where components are formed with a single conductive thread per pin).

The last possibility is the one explored most fully in the present work, because the others have been ruled out for various reasons.

- Solders used with electronic components are soft alloys of lead (*Pb*), tin (*Sn*), and sometimes silver (*Ag*). Such compounds are not suitable for use in applications in which they could potentially be in constant contact with a user’s body, because of their toxicity. Worse yet, while it is possible to solder components onto metallic organza and achieve (very) good electrical contact, the mechanical properties of the joint are unsuited to the flexure it will be subjected as an item of apparel.
- Conductive adhesives are better suited to this application than solder, because it is possible to envision adhesives that are non-toxic, highly conductive, highly durable, and moderately flexible to act as a “mechanical impedance match” between a flexible fabric substrate and a rigidly

packaged component. They remain an open possibility and should be the subject of further study.

- Stapled components are an interesting compromise, where a component lead grips a sewn conductive trace by being pressed into shape around it. When the substrate flexes, the trace is free to move within the clasp of the formed lead, forming a self-wiping contact at every junction between fabric circuitry and component pins. However, the dimensional rigidity of the component is a poor match for freedom of motion enjoyed by a fabric substrate, and the mismatch is likely to stretch open pins that have been formed into clasps and to accelerate wear and tear of the fabric substrate.

3.1 Flexible circuit substrates

Many methods already exist to fabricate circuits on flexible substrates. Most of these rely on the metallization of a flexible polymer substrate that can withstand the high temperatures of conventional soldering processes. Kapton film, for example, is one of the most commonly used flexible substrates, typically finding applications in cameras (where circuitry must fit intimately into available space in a small package), printers and portable computers (where a large number of connections must cross a hinge or other rotating joint), and non-planar antennas (where the antenna elements must have accurate dimensions yet be shaped to fit on conical forms).

When circuitry is designed for such a substrate, care must be taken to insure that in the final application mechanical stress will not be applied to the component-circuit solder joints. A consequence of this fact is that the parameters of motion of the flexible substrate must be well-understood at the time of design, or rather, that constraints must be imposed on the motion of conventional flexible substrates to ensure that they continue to work over their expected lifetime.

In this respect, conventional flexible substrates are flexible only in regions where components are not attached. Although many such substrates are well-characterised and have electrical properties highly suited to their use in forming electronic interconnects, their mechanical properties still leave something to be desired.

The case against conventional flexible substrates is summed up in the observation that they cannot be (non-destructively) crumpled the way that cloth can be.

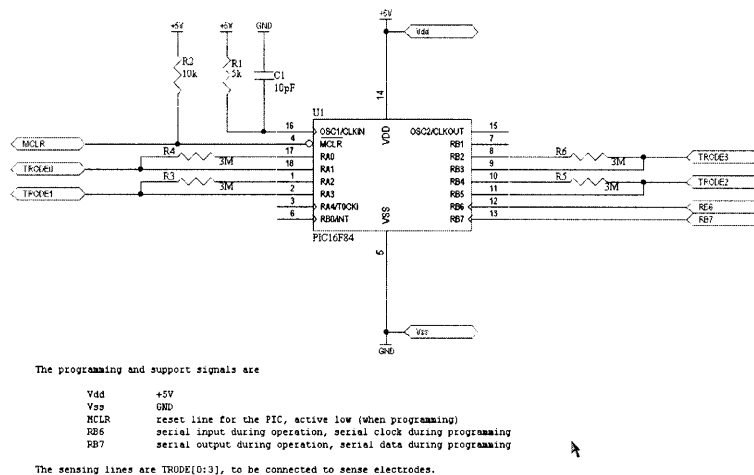


Figure 3.1: Schematic of the “flexible” multi-chip module.

3.2 Flexible multi-chip modules

An interesting exception to this rule might be the flexible multi-chip module (FMCM), an example of which is shown in Figure 3.2. This module is the result of a collaboration with workers at the Charles Stark Draper Laboratory to produce small, semi-flexible circuit elements that may be embedded in textile products to read sensors and report their status to a controlling host. The fabrication process of this module is similar to that developed by General Electric’s MCM-E/F group [Mar95].

The major part of this readout element is a PIC16F84 microcontroller IC, visible underneath the titanium-copper interconnections. In addition to the PIC16F84 there are four $3\text{ M}\Omega$ chip resistors and a 16 pF chip capacitor. A schematic for the MCM is shown in Figure 3.1, and is similar to the circuit used in the fabric keypad in its simplicity and choice of microcontroller.

In this case, the host attaches to the *RB7* and *RB6* pins, which form a full-duplex (RS232) serial data link operating at 9600 baud. The connections are formed by drilling out the centers of the pads and inserting a small pin or rivet to pressure-fit each pad to a mating e-brodered circuit trace. Power for the circuit is supplied across the *Vdd* and *Vss* pins, and e-brodered electrodes are connected to the *Trode[0-3]* pins to be capacitively read out, just as the original fabric keypad did.

The entire FMCM measures $0.600\text{ inch} \times 0.300\text{ inch}$, is 0.016 inch thick, and has a permissible bending radius of curvature of about 50 cm . To construct this FMCM, the constituent components were thinned to 0.006 inch by mechanical grinding and lapping of the rear face of each component. The components were then placed on a rigid substrate, a Kapton mask (also 0.006 inch thick) with

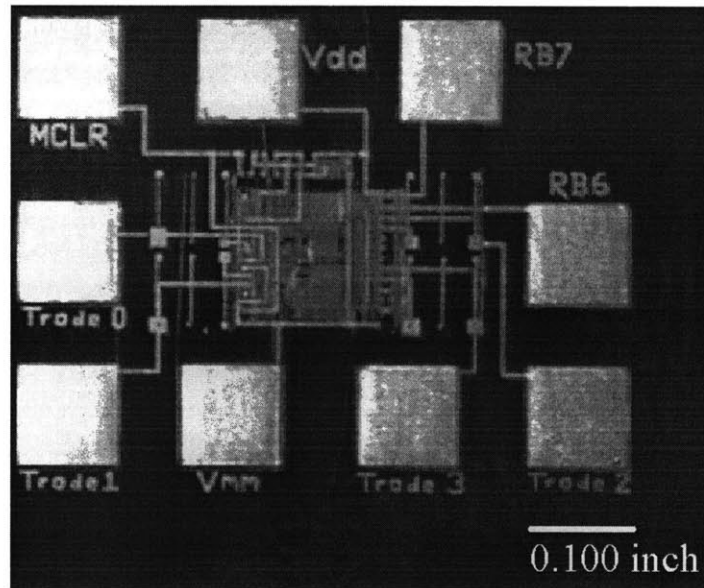


Figure 3.2: Micrograph of a “flexible” multi-chip module.

windows cut out to accommodate the chips was placed around the chips, and the Ti/Cu interconnect was deposited through a mask. This process extends to multilayer circuitry as well through the use of thin insulating layers. Finally, a protective Kapton cover is glued to the top of the circuit module, the module is removed from the rigid substrate, and a Kapton cover is glued to the bottom of the module, resulting in a completely encapsulated element.

3.3 Electrically active textiles

Many textiles have properties that suggest their use in electronic circuitry. There are all possible combinations of synthetic and natural fibers, coated or wrapped or spun with metallic fibers or conductive polymer fibers. In this section, some of these materials are identified and described.

3.3.1 Metallic silk organza

In its primary form, this is a finely woven silk fabric with a thin gold helix wrapped around each thread that runs along the weft of the weave (Figure 3.3.1). Believed to have originated in India, this sort of fabric has been produced for centuries, and appeared in Western fashion as early as the



Figure 3.3: Micrograph of metallic organza.

mid-18th century [Har97].

The warp of this fabric consists of parallel silk threads. Through this warp, the weft is woven with a silk thread that has been wrapped in a metal foil helix. This metallic thread is prepared just like cloth-core telephone wire, and is highly conductive. The silk fiber core has a high tensile strength and can withstand high temperatures, allowing the yarn to be sewn or embroidered with industrial machinery. The spacing between these fibers also permits them to be individually addressed, so a strip of this fabric can function like a ribbon cable. If a section of organza is subjected to shear, the cells formed by the weave also shear to form parallelograms, keeping the conductive fibers parallel and separated at all times.

Circuits fabricated on organza only need to be protected from folding contact with themselves, which can be accomplished by coating, supporting or backing the fabric with an insulating layer which can also be cloth. Also, circuits formed in this fashion have many degrees of flexibility (i.e. they can be wadded up), as compared to the single degree of flexibility that conventional substrates can provide.

In the microcontroller circuit shown in Figure 3.4, a PIC16C84 microcontroller and its supporting components are soldered directly onto a patch of metallic organza. This circuit uses the bidirectional

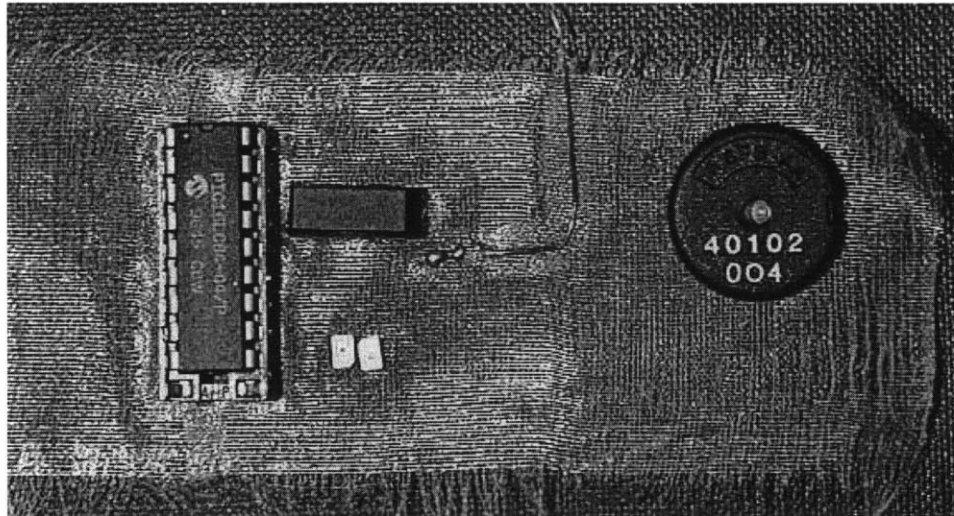


Figure 3.4: A fabric breadboard or “smartkerchief”.

I/O pins on the PIC to control LEDs and to sense touch along the length of the fabric, and uses audible feedback through a piezoelectric speaker to reinforce the sense of interaction. All of the components are soldered directly onto the surface of the metallic organza weave.

3.3.2 Conductive yarn

There are also conductive yarns manufactured specifically for producing filters for the processing of fine powders. These yarns have conductive and cloth fibers interspersed throughout. Varying the ratio of the two constituent fibers leads to differences in resistivity. These fibers can be sewn to create conductive traces and resistive elements.

3.4 Other interconnection strategies

3.4.1 Gripper snaps

Gripper snaps make excellent connectors between fabric and electronics. When a two-piece gripper snap is placed on fabric, the first piece has several metallic “teeth” which pierce the substrate and any conductors, making a wiping contact which scrapes off some surface contamination. The second piece is then pressure-formed with the first, cold-welding the teeth into the body of the snap. The

snap provides a robust electrical contact that allows subsystems snapped into clothing during use to be removed for washing.

3.5 Component packaging

By and large, integrated circuits with a large number of connections have pins (or other connections) placed on a regularly-spaced square grid. Connections are made to leads around the periphery of a rectangular package or (more densely) to pins or pads on the undersurface of a square package.

As the connection density increases, the leads themselves become shorter and narrower, as does the spacing between them. Of package styles currently in common use, the one with the greatest surface connection density is the ball grid array (BGA).

However, components destined for placement in fabric circuitry do not demand high connection densities, as a consequence of the design principles mentioned above.

3.5.1 The case for round packages

The packaging scheme proposed in the present work creates parts with threads leading out from the component. These threads are then attached to the embroidered circuit substrate by a covering stitch which holds each thread onto the substrate and electrically connects it to a particular sewn conductor.

Figure 3.5 illustrates components in square and round packages stitched onto a fabric substrate, while Figure 3.6 shows an actual stitched prototype component couched onto a fabric substrate.

For a given lead spacing at the periphery of a package (s) and a given package diameter (d) or side length (l) the maximum number of threads coming out of each type of package is found to be

$$n_{square} = 4\left(\frac{l}{s} - 1\right) \quad (3.1)$$

$$n_{round} = \frac{\pi d}{s} \quad (3.2)$$

For a given thread spacing around the periphery of the package, it's clear that for a large number of pins the square package will have a smaller diameter than the round package. But if the number of pins is less than 16 (or to be exact, $\frac{16}{4-\pi} - 4$), then the round package is will be a better use of available space.

The finest practical thread spacing achieved in the present work has been 0.050 inch (1.27 mm), which for 16 leads implies a package diameter of 0.254 inch (6.48 mm). It is practical to expect that a die of about half this width will fit into the package, which gives an upper bound of 10 mm^2 of single- or multi-chip real estate to work with in a typical sewable button package.

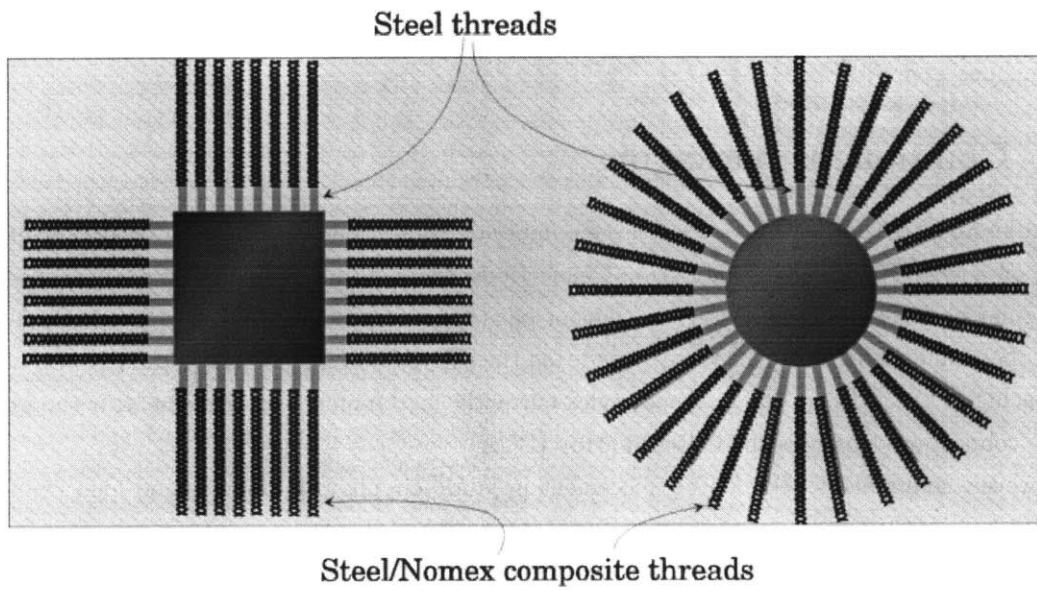


Figure 3.5: Square and round packages sewn onto fabric.

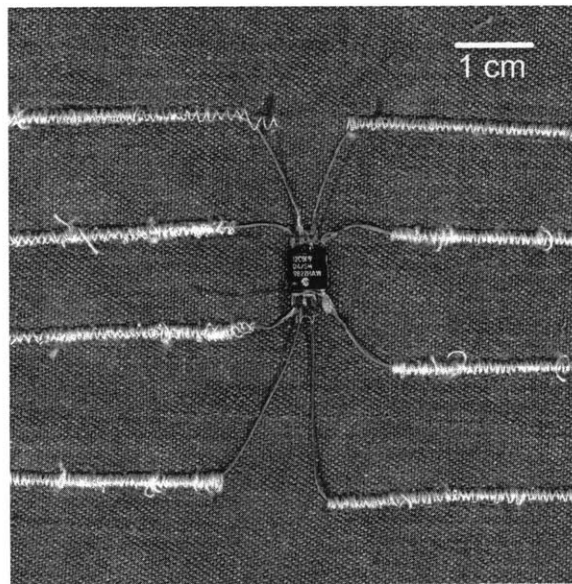


Figure 3.6: Prototype PTCC package sewn onto fabric.

3.6 The Plastic Threaded Chip Carrier

Now we look to the practical aspects of building packages like those that have been described so far. To manufacture threaded packages, we would start with a bare die, add a “threadframe”, and connect the near ends of the threadframe to the bondout pads of the die.

Figure 3.7 shows the possible construction of such a component package. Some details have been omitted for clarity, such as the metal base which helps to keep the die substrate at near-uniform temperature and electrical potential across its area.

Connections directly to the die are made in the usual way; fine gold wires are thermocompression-bonded to on one side to the die’s gold-plated “bondout pads” and on the other to the stubs of a conventional copper leadframe.

In most packages, the leadframe continues out of the plastic package and is ultimately soldered or cold-welded to external circuitry. But instead of having a solid external leadframe, the plastic threaded chip carrier (PTCC) has flexible, corrosion-resistant threads intended to connect to external circuitry.

In this case, the threads which leave the package are bundles of approximately one hundred continuous steel fibers, each about $5\ \mu\text{m}$ in diameter. These fiber bundles are microspotwelded to the leadframe stubs, and the entire structure is hermetically sealed in a plastic carrier.

3.6.1 Prototyping the PTCC

To test the proposition of a threaded chip carrier, prototypes must be built to evaluate their electrical, mechanical, and sartorial performance.

Since part of the PTCC’s assembly process is common to most IC packages, we can test the design principles behind the PTCC without building one completely from scratch. Leaded surface mount (SMT) IC packages already incorporate the die, the leadframe, and bonding wires in a sealed plastic carrier, and their electrical characteristics are well understood.

Prototypes have been built by starting with SMT packages, adding a few components directly to the leads, then microspotwelding steel fiber bundles to the leads. Finally, the threads are arranged in a radial pattern, and the entire assembly is encapsulated in an epoxy resin.

Packages made this way are larger than they need to be, but all of the new design principles involved are represented and open to refinement.

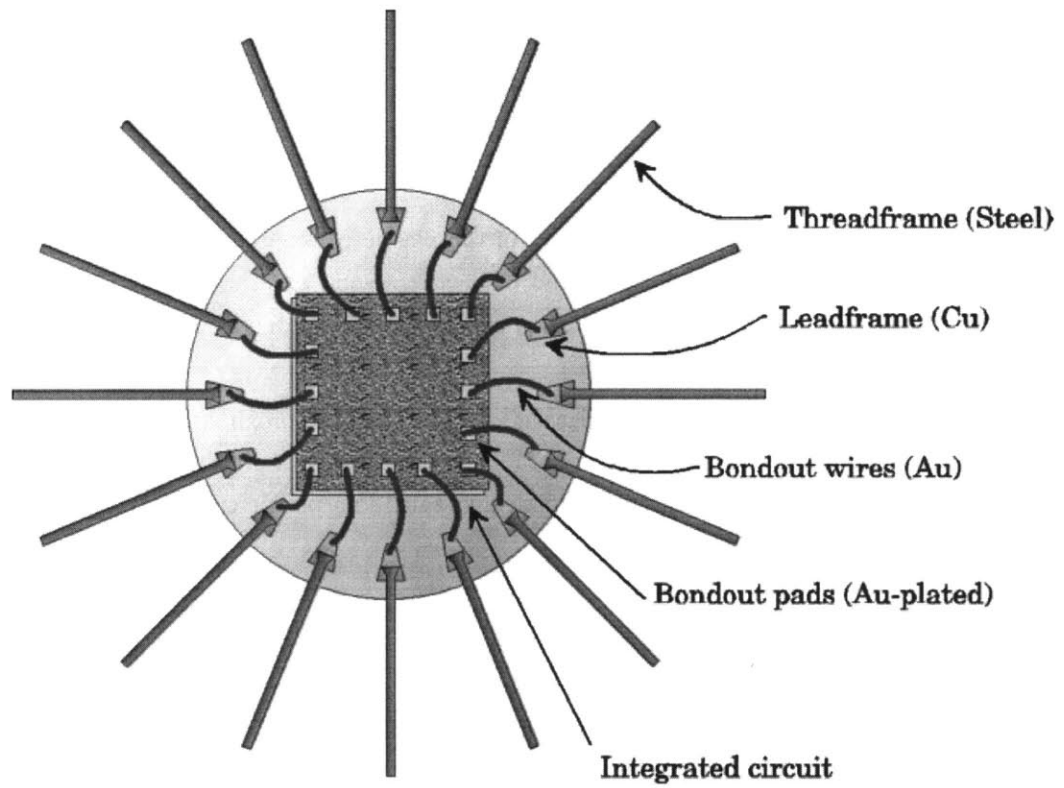


Figure 3.7: Internal structure of the Plastic Threaded Chip Carrier.

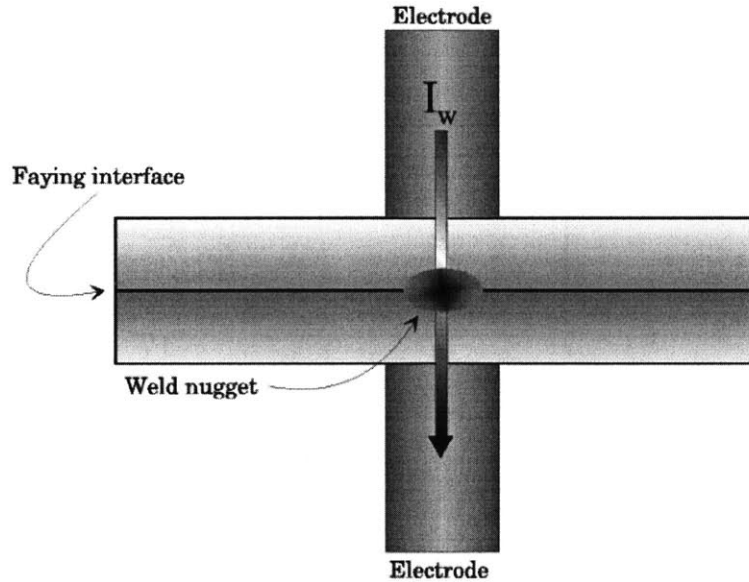


Figure 3.8: Schematic of weld junction.

3.6.2 Spot welding

Electrical or resistance welding was pioneered in the late 19th century by Elihu Thompson. Thompson took advantage of the fact that an electrical current evolves heat in any resistive material it passes through.

For constant-current welding the heat evolved, δQ , has the form

$$\delta Q = \frac{I_w^2 R}{A} t \quad (3.3)$$

where I_w is the weld current, R is the weld junction resistance, A is the weld junction area, and δt is the weld time.

The welding apparatus is arranged so that most of the weld power is dissipated at the faying interface, which is the junction of the materials to be welded. Inspection of the heat evolution equation reveals that for a given weld current and time, the weld heat increases directly as the junction resistance and inversely as the junction area. This suggests that the resistivity of the weld path should be concentrated at the weld junction, and that the junction area should be small.

Resistive heating is only part of the story, however. While a large area of contact is desirable between the electrodes and the workpiece to reduce heating at those points, a small electrode-workpiece

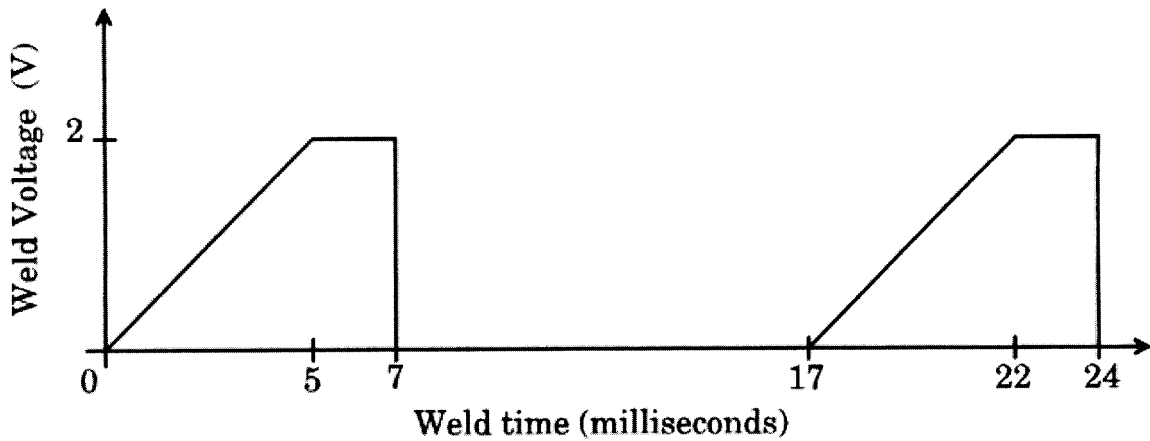


Figure 3.9: Schedule for welding $CuSnPb$ leadframe to steel fiber bundle.

junction area is preferred to reduce heat diffusion from the faying interface.

When two uniform, smooth pieces are welded together, the heating will be greatest along the center axis of the applied current, and a weld will form between the pieces along this axis. The heating is greatest along the center of this axis, because at that point the weld is most surrounded by heated material.

Different types of welds will arise from different levels of heating. The “coldest” welds form when the grain boundaries of the workpieces shift and flatten to form large interfaces. As heating increases, grain boundaries shift and dovetail, forming a complex interface. Still greater heating results in a breakdown of grain boundaries and a variable alloying of the materials making up the junction.

The leadframe used in this case was copper (Cu) plated with a tin-lead ($SnPb$) alloy. Diffusion alloying takes place at the $Cu - SnPb$ interface, and the result after several hours is a $Cu - CuSn - SnPb$ interface. This chemistry is common to the process of soldering with $SnPb$ alloys on Cu or $CuSn$ surfaces.

Weld schedules are often difficult to optimize, and depend crucially on many factors including electrode geometry, contact area, and electrode-workpiece metallurgy.

For the present work, weld schedules have been developed to maximize the pull strength of the thread-leadframe junction.

To prevent interference that would be caused by fraying of the end of the fiber bundle, it is first welded flat by one or two weld operations. The tab is then cut in the middle and the excess length is discarded. To weld this tab to the leadframe, the tinned copper leadframe ($Cu/SnPb$) is placed on

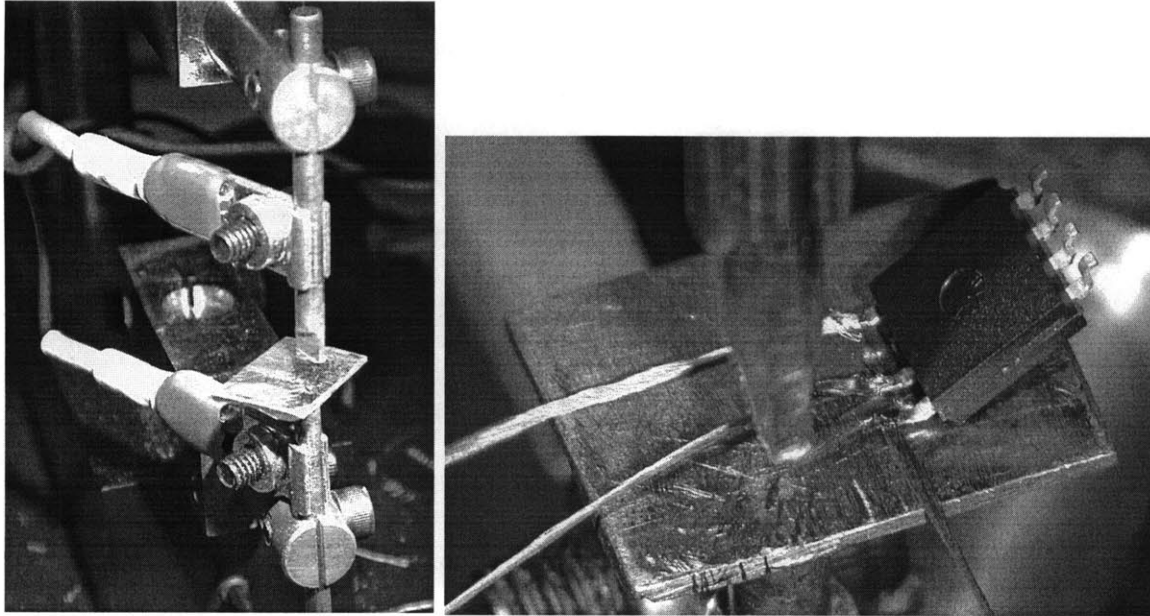


Figure 3.10: Welding electrode configuration.

the lower electrode, the steel fiber bundle tab is placed on top of the lead, and the upper *Cu* electrode is brought down to weld the tab to the lead at an applied force of 16 lbs.

Other weld schedules, geometries, and electrode materials have been tried with varying degrees of success. To minimize the number of tool changes necessary during weld process development, an electrode geometry has been chosen to simultaneously provide a good mix of thermal, electrical, and mechanical properties, and is shown in Figure 3.10.

The upper electrode is formed from $\frac{1}{8}$ inch diameter copper rod stock, with the end ground to a $\frac{1}{8}$ inch \times $\frac{1}{16}$ inch rectangular cross section. The lower electrode is formed of $\frac{1}{8}$ inch copper rod with a 1 inch \times $\frac{3}{4}$ inch \times $\frac{1}{16}$ inch copper “anvil” soldered on top. The anvil provide a thermal sink which draws excess heat away from the weld, decreasing the likelihood of electrode sticking and promoting even heating of the weld joint.

3.6.3 Weld evaluation

The mean shear strength of a weld formed according to the above procedure is 36.8 *N* ($\sigma = 4.3$ *N*, $n = 9$), determined by pull testing. Welds with a pull strength less than 10 *N* were discarded (3 out of 12 samples tested) as resulting from poor process control or weld contamination. During these

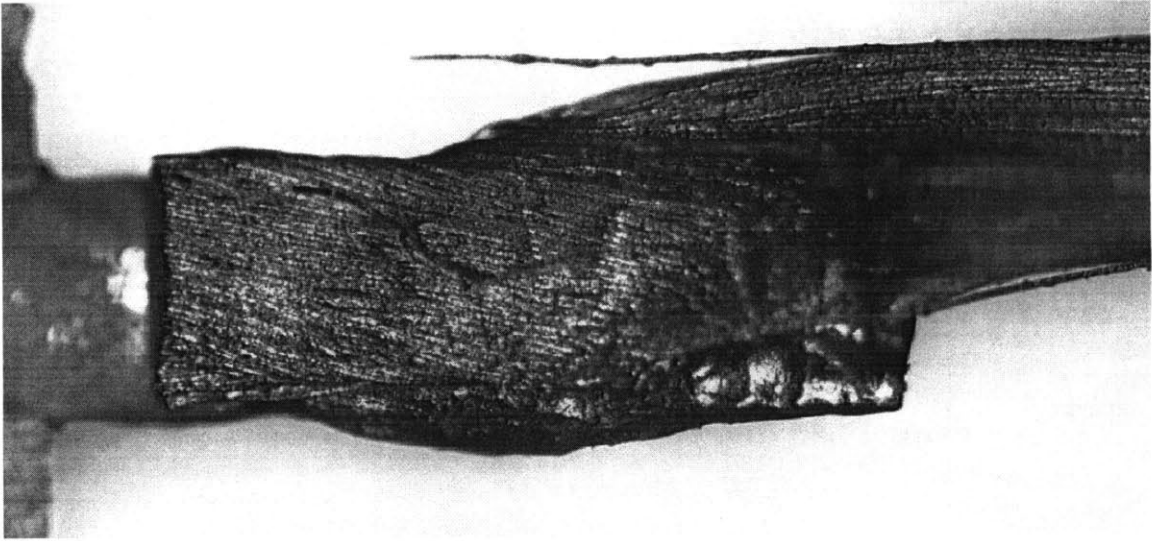


Figure 3.11: Top view of a stainless steel thread welded to $Cu/CuSnPb$ lead.

tests, most failures occurred at the point where the leadframe entered the plastic package, not at the thread-leadframe weld itself.

A pull strength of $36.8N$ (or equivalently, $8.25lb$) is more than sufficient for the application described above, *e.g.* that of welding steel threads to leadframe stubs prior to encapsulation in a hermetically-sealed plastic package. It should also suffice for the purposes of prototyping such packages, where steel yarn is welded to the leads of an existing package and the resulting construction is encapsulated in epoxy. In both of these applications, the pull strength of the original weld is augmented by the threads' adhesion to the package encapsulant, and the compressive constraint the package itself imposes on the weld junction.

3.7 E-broidered circuitry

Once the components are packaged, they are sewn onto an insulating fabric substrate with embroidered conductive traces. The traces themselves are sewn in a composite yarn, spun from a mixture of 95% Nomex fibers and 5% steel fibers, where the average fiber length is about 3 *cm*. The fine structure of this yarn is shown in two views in Figure 3.13.

A composite yarn of this sort is used because the all-steel thread used to form component leads cannot be sewn by ordinary processes, comprising as it does a bundle of some hundred continuously-

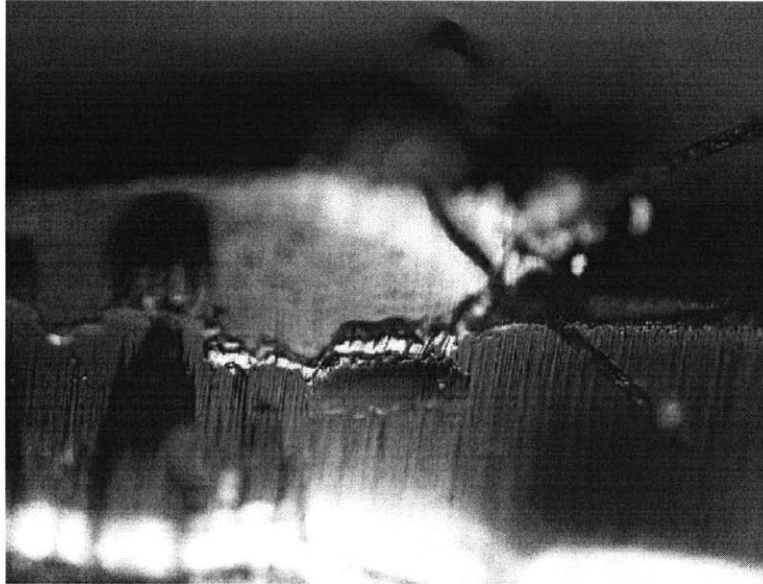


Figure 3.12: Optical micrograph of *Cu*-steel weld junction.

drawn steel fibers. As a thread is sewn into fabric, tensions are imposed unevenly over its cross section. Threads spun from short (staple) fibers are able to accommodate this variation in tension by stretching more on one side than the other, and manage to rebalance the tension after a short stretch of thread. Uneven stresses propagate only over short distances in staple yarns.

Contrast this with the continuously drawn steel fiber bundle, in which individual fibers are meters (rather than centimeters) long. Stresses are no longer local to a small region of sewing, and tensions rapidly mount without resolution, resulting in bunching of the thread as it feeds through the sewing machine. Some of the bunched threads will eventually twist into an obstruction that will not pass through the sewing needle's eye, and the entire process will grind to a halt.

An alternative to the continuously drawn steel fiber bundle might be an all-steel yarn spun from short steel fibers. Care would have to be taken to choose the length of the fibers to balance the effects of stress propagation over long distances against the tendency of the fibers to not interlock over short distances because of their stiffness.

The particular composite yarn used here is known as Bekintex, and is manufactured by Bekaert Fibre Technologies. Bekaert Fibre produces a variety of composite textiles which incorporate finely drawn metal fibers to obtain particular bulk properties of electrical conductivity, thermal conductivity, surface finish, and tensile strength.

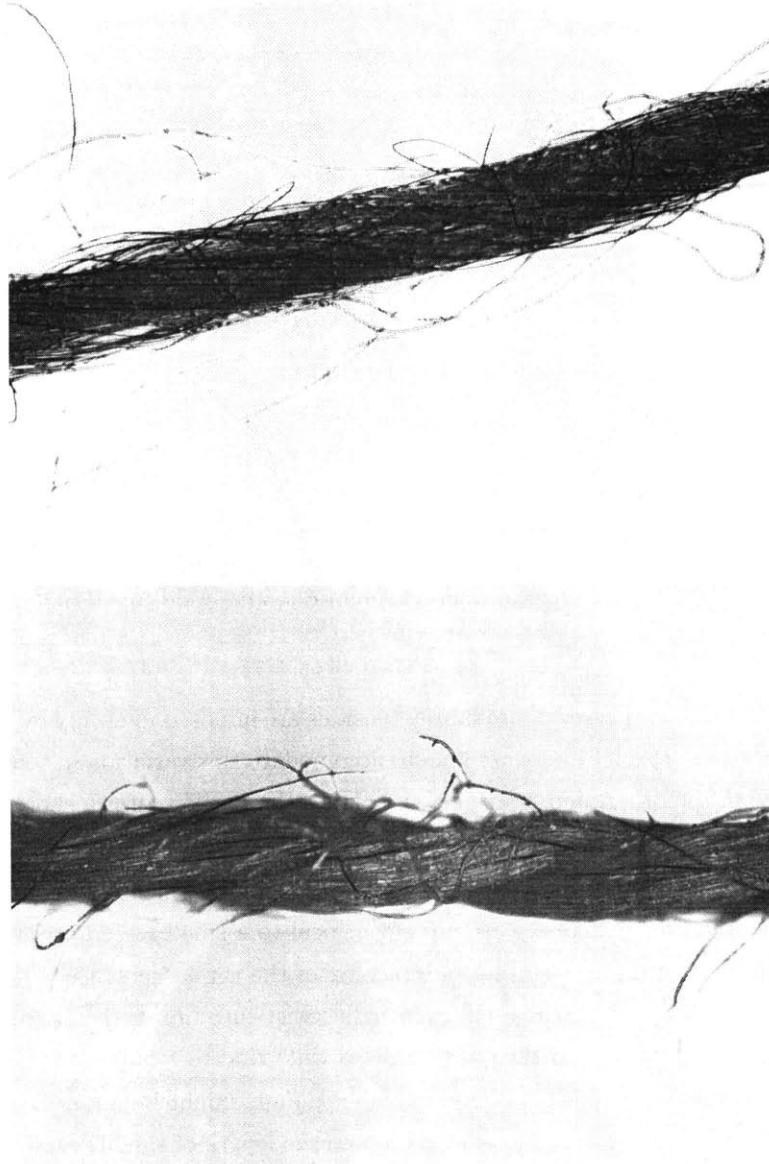


Figure 3.13: Optical micrograph of stainless/Nomex composite thread (Bekintex 50/2). $20\ \mu\text{m}$ stainless steel fibers are visible as dark lines in the lighter-colored Nomex matrix that forms the bulk of the yarn. This two-ply yarn is approximately 0.010 inch in diameter.

Bekaert credits a 1936 U.S. Patent [Eve36] with the original description of the process whereby a bundle of fine metal fibers may be drawn continuously and simultaneously from source metals. Bundle drawn fibers can be produced in many different morphologies by translating the polymer-based methods of the synthetic textile industry into their metallurgical equivalents. Continuous and broken bundles, cut fibers, spun yarns, threads, *etc.* are all produced in this way.

The fibers themselves are available in diameters from $100\mu m$ down to $12\mu m$, and as far down as to $2\mu m$ for very metallurgically clean (free of large inclusions) alloys. Alloys commonly used in these processes include Stainless 316L steel, Stainless 302 steel, Inconel 601, Nichrome, FeCr alloy, and Titanium [Tec97]. These 100% metal fiber yarns are used in knitting, weaving, needlepunching, and braiding processes, with sewing and embroidery being notable omissions from the list.

Bekaert cites the primary applications for these materials as being filter media, antistatic textiles, heat-resistant textiles, burners, and conductive plastics. Their product range includes 100% metal fiber products as well as blends of metal fibers combined with natural or man-made fibers [Tec97].

Steel threads were chosen for their strength, resistance to corrosion, biological inertness, and ready availability in textile form at low cost. The major drawback of using steel and steel-composite threads is the difficulty involved in attaching them to existing electronic components.

For instance, we might wish to attach steel threads directly to the bondout pads of an IC die without using an intermediate *Cu/CuSnPb* leadframe. Unfortunately this is impractical for many reasons.

First of all, steel threads are too strong to be thermocompression bonded to metallized pads on a crystalline silicon substrate without causing the substrate to crack [KRT89]. More important however is the metallurgy of bondout pads on silicon chips.

Most ICs use *Al* metallization for on-chip interconnections, as well as for bondout pads. Nearly all pad-to-leadframe wirebonds take place using *Au* or *Al* wire (actually *Al*99% + *Si*1%). *Au* is used more frequently because it requires less pre-forming during the wirebonding procedure and because it forms less of a surface oxide layer than *Al* wire does. Bondout pads are also usually plated with *Au* (over the *Al*) to promote good wire-to-pad bonding.

The *Au – Al* interface does however represent a compromise between the need to achieve a strong wirebond and the need to develop a process insensitive to contamination. The most common reliability problem in wirebonding results from the formation of intermetallic compounds at the *Au – Al* interface, known by their color as “purple plague” (*AuAl*₂) or “white plague” (*Au*₅*Al*₂). These compounds are very brittle compared to the metallic *Au* and *Al* that surround them. As a consequence of this brittleness, wire flexing (as a result of vibration or thermal cycling) more easily induces metal fatigue and stress cracks [KRT89]. These and other intermetallic compounds and alloys deleterious to the strength of the *Au – Al* interface are known to form more readily in the presence of all of the metals

that Bekaert produces in textile form. It is therefore essential to ensure the purity of the gold used in plating bondout pads, as well as the purity of gold wire used to bond to the gold plated pad.

3.7.1 E-broidered composites

One very useful technique which arises from the attachment technique used here (couching the thread-frame to the fabric with an e-broidered trace) is that of on-the-fly stitching of composite traces. Over the run of the trace which couches the steel thread, the trace's conductivity will be approximately that of the steel thread itself (slightly higher due to the parallel conductivity of the e-broidered trace). There is a difference in conductivity of almost two orders of magnitude between the stainless thread and the stitched Bekintex composite yarn, with typical resistances of $1\Omega/cm$ and $100\Omega/cm$ respectively.

This immediately implies that low-tolerance resistors can be constructed simply by patterned stitching of known lengths of Bekintex and Bekinox. It also leads to the more general notion of a composite trace or sensor generated by mixing materials together by means of stitching.

3.8 Implementation of the MIDI Jacket keypad

The MIDI Jacket keypad was built using a PIC microcontroller to perform capacitive measurements of connections to sewn electrodes on a denim substrate. The measurements were implemented almost entirely in software, as an exercise in developing electric field sensors using a minimum of hardware, but this minimalism also benefits the fabric circuit designer by reducing the number of components (and hence interconnections) that must be incorporated into a circuit.

The keypad has been shown before, in Figure 1.4, while its circuitry and software are detailed in the Appendix. A washable electrode array was sewn into denim using Bekintex thread, in the pattern of a telephone keypad with traces leading from the symbols to an array of connection pads intended to mate with a conventional circuit board.

3.9 Multilayer circuit construction

The leap from single layers of e-broidery to multi-layer circuitry is motivated not only by the desire to develop non-trivial interconnects and circuitry, but also by the need to control the outward appearance of fabric interfaces. Fabric-based user interfaces should at large display only devices and symbols relevant to user interaction, and not the underlying complexity of their implementation.

leads to the final step of stacking fabric layers with and adding an inter-layer stitched interconnect. The way this is done with conventional circuit boards is to print each layer in copper on one side of a thin substrate, then to laminate the strata together, drilling holes as appropriate and plating with copper to interconnect traces on different strata. The inter-layer connections are commonly referred to as vias, and the metallurgy and chemistry of this process is messy and expensive.

To do the same thing in fabric, however, we have an easier time. Each layer of the interconnect is e-broidered onto its own fabric substrate. Then insulating fabric layers are placed between the e-broidered layers, with vias stitched between layers using more conductive thread.

Suppose we want to build an interface which has only buttons on its top surface. How can this be done?

First, we e-broider a plane with nothing but the interface objects sewn in. The next layer is simply a thin, sturdy insulating layer, followed by another layer with e-broidered circuitry. This second layer of circuitry includes pads which line up with the interface symbols on the top layer. The next layer is again an insulator, and the last e-broidered layer includes the components which have had their leads couched onto the fabric substrate connecting to traces, some of which include pads to connect with other layers. Finally, the layers are assembled and stitched together at points where signals must cross between layers.

To enhance the final appearance, the topmost layer is the last one to be quilted onto the remainder. This avoids sewing vias through the top layer for any connections other than those required by the top layer. The principle is illustrated in Figure 3.14.

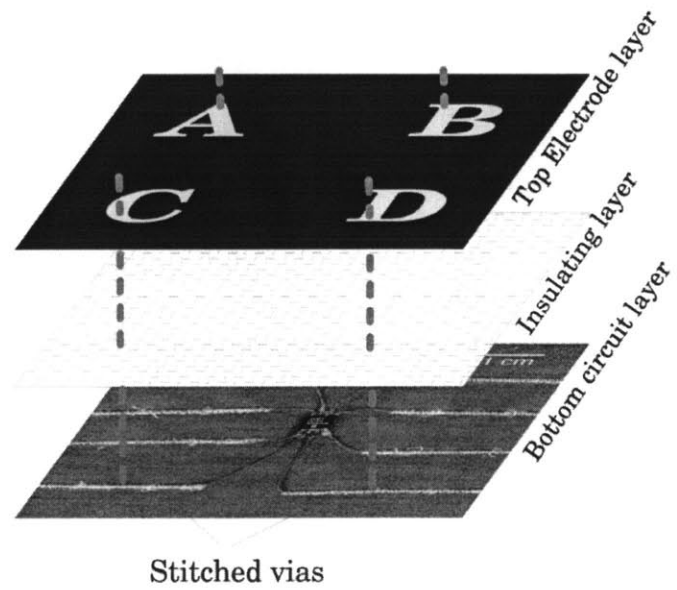


Figure 3.14: Illustration of the layers making up the No Soap Radio.

Chapter 4

Future work

What remains to be done? The present work collects a year’s worth of observations and experience with fabric circuitry in a desire to point the way toward scalable processes for the manufacture of washable computing systems. It is expected that the primary uses of such systems will be as interfaces to more commonplace devices, such as pagers, phones, and wearable computers.

The two biggest questions that arise in connection with this work are “Can you wash it?” and “What if you want to wear a different outfit?” The present work addresses the first question directly, and the second indirectly.

Regarding washability, e-broidery will not currently function when it is wet, but it works fine once it is dry again. Whether the wetting agent is rain, isopropyl or ethyl alcohol, or one of the common dry-cleaning fluids, once the e-broidered circuit is dry it functions as specified. Although rigorous testing remains to be done, the early indications are very promising. Even human sweat (a mildly corrosive saline condensate) does not appear to harm e-broidery even over many cycles of exposure, while it does in fact corrode conventional tinned-copper printed circuitry. Moreover, if a conventional circuit board gets wet during operation, solder will be seen to dissolve sufficiently into the water to form aggregates along electric field lines in the water, causing shorts to form. E-broidery does not appear to be subject to this mechanism, although further testing of the steel threads is on the agenda.

The threadframe prototypes have been surprisingly robust, and further work should be done to characterize their integrity over multiple stress cycles. Since the stainless threads used in this work are composed of many long, thin fibers, we would like to know how often an individual fiber is likely to break when the thread or the entire structure holding it is subject to stress.

As for the issue of changing outfits, this is an open research question that might be best addressed by ongoing efforts into building environments that support mobile agents in embedded applications.

One vision of the washable computer of the near future is a jacket with many small computing elements distributed throughout, all interconnected by a stitched network. Each node would provide sensing in its physical locale, some storage for a distributed, redundant database, and a computing element capable of executing small mobile applications. A jacket or vest as a whole could easily host a hundred such small nodes (an added weight of about 50 grams or 2 ounces), and be powered all day by a battery slipped into a pocket. When you hang up your jacket, it joins the closet network and synchronizes itself with other articles of clothing and a host computer. The scenario is easy to construct given an embeddable substrate for mobile code.

Clothing is also interesting because it can augment intrabody signalling [Zim95] and power transduction [PRPG97] techniques. The capacitive links to the human body that these depend on are limited by the total area of capacitive contact available at each link. By transmitting power harvested in the shoes through the body to a large receiving electrode in the jacket, efficiencies should be realized that are sufficient to replace the need for batteries.

Much work remains to be done in replacing Bekinox and Bekintex with composite textiles designed to withstand corrosion, to allow easy welding and sewing, and to have higher conductivity. There is also another way that batteries might be replaced, by collecting the triboelectric currents that cause charge migration and “static cling”, but this will require careful attention to the topology and chemistry of weaves intended to generate power.

Many interesting proposals to make wearable information infrastructures suggest the application of highly redundant optical fiber networks in clothing. We do not see e-broidery as a replacement for optical fibers, but rather as a complement. E-broidery excels at certain things when compared to optical fiber, but fiber also has its shining points. Many successful systems are likely to stem from a clever combination of the two.

Apart from e-broidery, electrospinning [RC96] may also be a viable way to place electronic textiles on a substrate. It may turn out to be a good way to bond components to e-broidered circuitry, for example, by allowing one to deposit polymer meshes of different properties onto an existing cloth substrate. In electrospinning, the force of electrostatic repulsion generates a fine, splaying jet of polymer (from a droplet of melt or solution) that forms a mesh at an electrically grounded collection surface. This may be better than spot-welding or applying conductive adhesives because electrospinning is guided by electric fields, so the spinning process will presumably extend previously-deposited conducting paths and insure their continuity.

Electrospinning is also important because it deposits a permeable mesh, rather than a solid, impermeable layer (as would silk-screening or printing). A mesh spun from a given material often has a better strength-to-weight ratio than the material would have in bulk. The mesh also increases the material’s flexibility and breathability as well as its total surface area. By electrospinning with

different materials, one can build multi-layer electronic structures much as printing would allow.

Wearable displays could be formed by combined electronic ink with an electronic textile substrate [Jac97]. One open challenge is to make a thread that encapsulates electronic ink and can be sewn into or onto ordinary fabric. There are other possibilities to be explored as well, including large-pixel displays and multi-segment indicator arrays.

Large sensor surfaces are worth exploring. Textiles range in scale from small pieces of needlepoint ($10^{-4}m^2$) to tapestries and carpets (10^2m^2), with a corresponding range of feature size. This indicates that textile processes may be well-suited to the production of large, flexible sensing surfaces that conform to any underlying shape.

Finally, a large interesting area of work to follow will be in using these techniques to construct meaningful and compelling user interfaces and applications. Some of the more obvious applications include clothing that can use electric field measurements to determine its fit and shape, keyboards invisibly and comfortably stitched into sleeves, cuffs, and pockets, and hybrid structures that incorporate electronic ink into clothing to create flexible displays. The work so far has been exciting and promising, and I hope it will continue to grow.

Appendix A

MIDI Jacket keypad implementation

Briefly, the increase in a sewn electrode's capacitance due to finger contact is measured in terms of the time taken to charge the electrode/finger capacitance from zero to the switching threshold voltage of a CMOS logic buffer ($\approx \frac{1}{2}V_{CC}$).

One variant of this measurement strategy is illustrated in Figure A.1. The main advantages of this technique are that it is a time measurement, it provides a dynamic range of 10^4 , and it can be implemented with little more than standard CMOS logic. Not only can direct contact be measured, but the dynamic range also permits non-contact sensing to occur over a distance roughly twice as far away from an electrode as it is wide.

There are also a number of ways to refine the dynamic range of the measurement in software. For example, the measurement granularity of the charging interval is ultimately limited by the length of the software loop which performs the time count. Normally one might discharge the capacitors, then turn on the charging signal and count the number of iterations of a fast loop that elapse before the capacitor voltage exceeds the input high threshold.

Instead of letting the capacitor charge continuously, however, the software that times its charging curve can also mete out a smaller pulse of current on each iteration, which means that each tick of the counter corresponds to a smaller portion of the charging curve.

For the MIDI jacket keypad, such careful measurements were not employed. Instead, a simpler procedure was employed, with each electrode permanently connected to analog ground through a $1M\Omega$ resistor, to allow the measurement of an electrode's discharge curve. For later systems, however (including the flexible MCM and the washable keypad), the technique described above was employed.

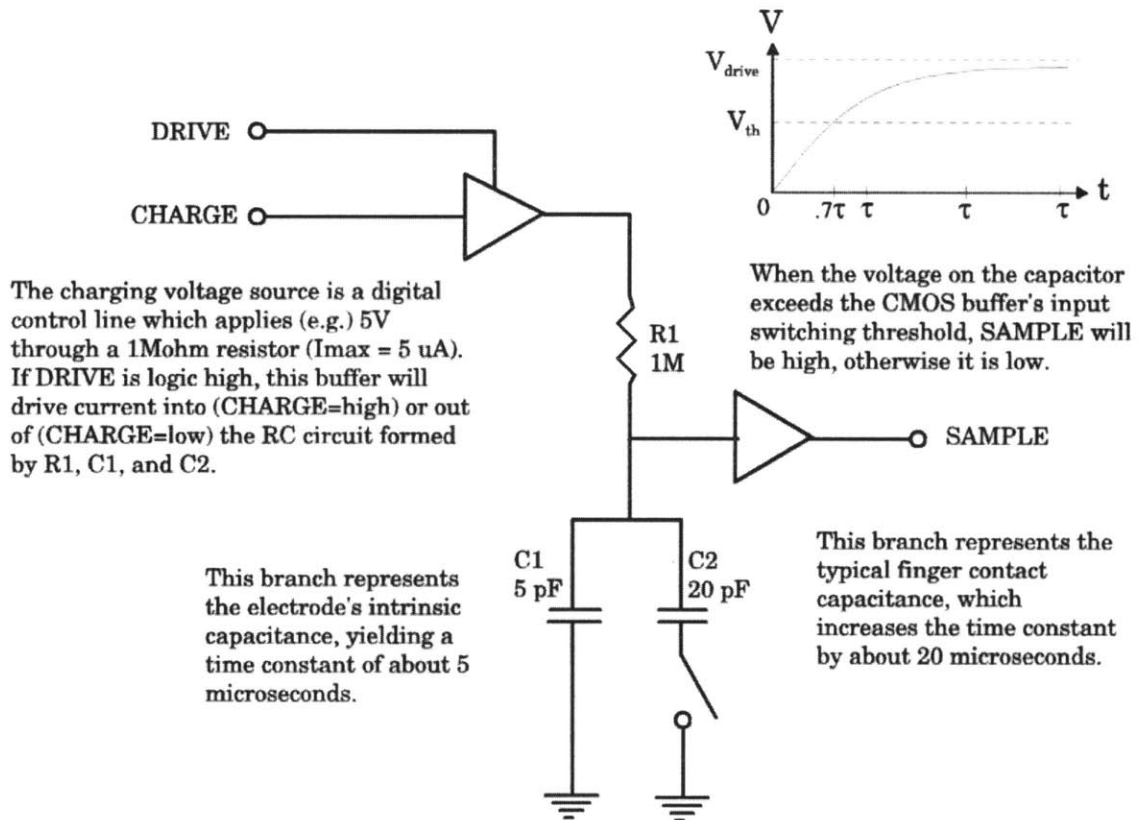


Figure A.1: Schematic illustration of RC charging capacitance measurement.

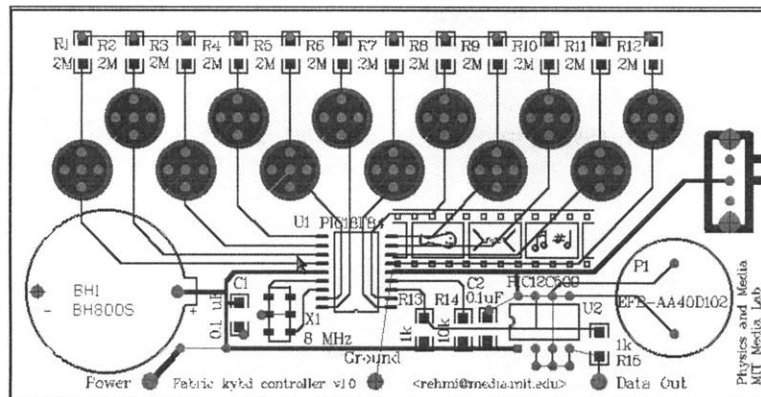


Figure A.2: PCB layout of the fabric keypad readout board.

```

1 ; Code for the fabric keypad controller
2 ;
3 ; Rehmi Post <rehmi@media.mit.edu> 10/11/97
4 ;
5 ; Set device to PIC16F84, hex radix for all calculations
6     processor      16F84
7     radix          hex
8     include        "p16f84.inc"
9 ;
10 ; Defines for PIC16F84
11 ;
12 PORTA             equ    05      ; Register mapping for PORTA
13 PORTB             equ    06      ; Register mapping for PORTB
14 STATUS           equ    03      ; Status register
15 CARRY             equ    0       ; Carry bit in status register
16 ZERO             equ    2       ; Zero bit in status register
17 SAME             equ    1       ; Destination register same as source reg
18 RPO              equ    5       ; Bit 5 is RPO
19 TRISA            equ    85      ; TRISA in Bank 1
20 TRISB            equ    86      ; TRISB in Bank 1
21 RBPU            equ    7       ; /RBPU in OPTION

```

```

22  OPT          equ    81      ; OPTION register in Bank 1
23  w           equ    0      ; Put back into w
24  RTCC        equ    1      ; RTCC in Bank 0
25  ;
26      variable RS232_INVERT=0 ; mimic a level shifter inversion
27  ;
28      variable      n=1
29  ;
30  ;
31  clockrate    equ    .8000000 ;define clock rate here
32  baudrate     equ    .1200     ;define baud rate here
33  ;
34  fclk equ     clockrate/4
35  ;
36  ;The value baudconst must be a 8 bit value only
37  baudconst    equ    ((fclk/baudrate)/8 - 2)
38      if baudconst > 255
39          error  "baudconst doesn't fit in 8 bits!"
40      endif
41  ;
42  ; Program Variables
43  ;
44  i           equ    0x0C
45  j           equ    0x0D
46  k           equ    0x0E
47  l           equ    0x0F
48  count       equ    0x1C     ;used to count tx and rx data bits
49  txreg       equ    0x1D     ;used as temp transmit register
50  delay       equ    0x1F     ;used to time the baud rate.
51  nothings    equ    0x2C     ;number of nulls to send in here
52  sent_char   equ    0x2D
53  ;
54  #define _tx  PORTA,4
55  ;
56

```



```

57 declblk macro  basename,baseaddr,num
58     local  m=0
59     while  m < num
60 basename#v(m) equ  baseaddr+m
61 m++
62     endw
63     endm
64
65     declblk cnt,010,D'12'
66     declblk off,020,D'12'
67
68 #define PAGE_0 bcf  STATUS,RP0
69 #define PAGE_1 bsf  STATUS,RP0
70
71     org  0x0000      ; Reset vector
72     goto 0x0005      ; Jump to Main
73
74     org  0x0005      ; Start here
75 Main
76     ; set up the ports to be all inputs except for PORTA,4 (_tx)
77     PAGE_1
78     movlw 00FH
79     movwf PORTA
80     movlw 0FFH
81     movwf PORTB
82     PAGE_0
83     ; establish baseline measurements of capacitance
84     call  baseline
85 again
86     call  measure      ; measure discharge times
87     call  adjust      ; adjust them accordingly
88     call  output      ; output data accordingly
89     call  pause       ; wait a little while
90     goto  again      ; do it again
91

```

```
92
93  pause
94      movlw  .26          ; 10 ms
95      movwf  j
96      clrf   i
97  pause0 decfsz i,f      ; 384 us / iteration on i
98      goto  pause0
99      decfsz j,f
100     goto  pause0
101     return
102
103
104  charge ; precharge all of the output lines and their associated electrodes
105     movlw  OFF
106     movwf  PORTB
107     if    RS232_INVERT
108         movlw  00F
109     else
110         movlw  01F
111     endif
112     movwf  PORTA
113     ; now drive all outputs
114     PAGE_1
115     movlw  000H
116     movwf  PORTA
117     movlw  000H
118     movwf  PORTB
119     PAGE_0
120     nop
121     nop
122     nop
123     nop
124     nop
125     nop
126     nop
```

```
127         nop
128         ; set up the ports to be all inputs except for PORTA,4 (_tx)
129         PAGE_1
130         movlw    00FH
131         movwf    PORTA
132         movlw    0FFH
133         movwf    PORTB
134         PAGE_0
135         return
136
137  avg      macro   in,cum
138           movf    in,w
139           addwf   cum,f
140           rrf     cum,f
141         endm
142
143
144  sample  macro   port,bit,dest,label
145           clr     dest
146           call   charge
147           movlw  03F
148           movwf  k
149  label
150           btfsc  port,bit
151           incf   dest,f
152           btfsc  port,bit
153           incf   dest,f
154           btfsc  port,bit
155           incf   dest,f
156           btfsc  port,bit
157           incf   dest,f
158           decfsz k,f
159           goto   label
160         endm
161
```

```
162  measure                               ; Measure contact at pads.
163      sample  PORTA,0,cnt0,10
164      sample  PORTB,0,cnt1,11
165      sample  PORTA,3,cnt2,12
166      sample  PORTB,4,cnt3,13
167      sample  PORTB,1,cnt4,14
168      sample  PORTA,2,cnt5,15
169      sample  PORTB,5,cnt6,16
170      sample  PORTB,2,cnt7,17
171      sample  PORTA,1,cnt8,18
172      sample  PORTB,6,cnt9,19
173      sample  PORTB,3,cnt10,1A
174      sample  PORTB,7,cnt11,1B
175      return
176
177  baseline
178      movlw   080
179      movwf   j
180
181  clear  macro
182      local  m=0
183      while m < D'12'
184          clrf   off#v(m)
185  m++
186      endw
187      endm
188      clear
189
190  baseloop
191      call   measure
192
193      avg    cnt0,off0
194      avg    cnt1,off1
195      avg    cnt2,off2
196      avg    cnt3,off3
```

```
197         avg     cnt4,off4
198         avg     cnt5,off5
199         avg     cnt6,off6
200         avg     cnt7,off7
201         avg     cnt8,off8
202         avg     cnt9,off9
203         avg     cnt10,off10
204         avg     cnt11,off11
205
206         decfsz  j,f
207         goto   baseloop
208
209  bolster macro  reg
210             clrc
211             rlf  reg,f
212             endm
213
214             ; make all the offsets twice their original value.
215
216             bolster off0
217             bolster off1
218             bolster off2
219             bolster off3
220             bolster off4
221             bolster off5
222             bolster off6
223             bolster off7
224             bolster off8
225             bolster off9
226             bolster off10
227             bolster off11
228
229             return
230
231  adjust
```

```
232         return
233
234
235 flagkey macro  val,thresh,char
236             movlw  char
237             movwf  txreg
238             movf   thresh,w
239             subwf  val,w
240             skpnc
241             call  out_char
242         endm
243
244 output
245         clrfsent_char
246         flagkey cnt0,off0,.17 ;'0'
247         flagkey cnt1,off1,.8  ;'1'
248         flagkey cnt2,off2,.12 ;'2'
249         flagkey cnt3,off3,.15 ;'3'
250         flagkey cnt4,off4,.6   ;'4'
251         flagkey cnt5,off5,.10  ;'5'
252         flagkey cnt6,off6,.13  ;'6'
253         flagkey cnt7,off7,.1   ;'7'
254         flagkey cnt8,off8,.5   ;'8'
255         flagkey cnt9,off9,.8   ;'9'
256         flagkey cnt10,off10,.13 ;'*'
257         flagkey cnt11,off11,.20 ;'#'
258         ; in which case we send the number of nothings remaining
259         call  sendnothings
260         return
261
262 sendnothings
263         movf  sent_char,f
264         skpz
265         return ; return if we sent a char
266 nothingloop
```

```
267     movf    nothings,f
268     skpnz
269     return          ; return if no nothings to send
270     movlw   0
271     movwf   txreg
272     call    out_byte
273     decf    nothings,f
274     goto    nothingloop
275
276 out_char
277     movlw   .1
278     movwf   nothings
279     movf    txreg,w
280     movwf   sent_char
281 out_byte
282 ;     movwf   txreg
283     if RS232_INVERT
284         bsf    _tx          ;send start bit
285     else
286         bcf    _tx          ;send start bit
287     endif
288     movlw   baudconst
289     movwf   delay
290     movlw   .9
291     movwf   count
292 txbaudwait
293     nop
294     nop
295     nop
296     nop
297     nop
298     decfsz  delay, F
299     goto    txbaudwait
300     movlw   baudconst
301     movwf   delay
```

```
302         decfsz count, F
303         goto  SendNextBit
304         movlw  .9
305         movwf  count
306     if RS232_INVERT
307         bcf    _tx            ;send stop bit
308     else
309         bsf    _tx            ;send stop bit
310     endif
311     if 1
312 stopwait
313         nop
314         nop
315         nop
316         nop
317         nop
318         decfsz delay, F
319         goto  stopwait
320     endif
321         call  intercharpause
322         return
323
324 SendNextBit
325         rrf    txreg, F
326         btfss STATUS,C
327         goto  Setlo
328     if RS232_INVERT
329         bcf    _tx            ;send one bit
330     else
331         bsf    _tx            ;send one bit
332     endif
333         goto  txbaudwait
334 Setlo
335     if RS232_INVERT
336         bsf    _tx            ;send zero bit
```



```
337     else
338         bcf     _tx         ;send zero bit
339     endif
340         goto   txbaudwait
341
342
343 intercharpause
344     call   pause
345     call   pause
346     call   pause
347     call   pause
348     call   pause
349     call   pause
350     call   pause
351     call   pause
352     call   pause
353     call   pause
354     return
355
356     END
```


Appendix B

Washable keypad implementation

The washable keypad is also referred to below as the No-Soap Radio, as it was designed to be dry-cleanable and to transmit data via near-field electrostatic coupling. When the user touches a symbol on the attached e-broidered keypad, the microcontroller capacitively sense the contact and then transmits a particular tone (depending on which symbol was touched) by FM-modulating a 455 kHz square-wave carrier in software and presenting the resulting signal on all of the keys connected to the circuit (Pad0 through Pad3). Two pins are reserved for intra-sensor communications to build larger keypads which communicate with each other and a host processor over a serial bus.

The only external components used in this circuit are two $2M\Omega$ resistors, which allow the charging time measurement to be done in software. These are included in lieu of resistors implemented directly on the CMOS process die, and are soldered onto the PIC12C509 leadframe before or after steel threads are microspotwelded onto the leads. The resistor packages are of the surface-mount 0402 type, with dimensions $0.040in \times 0.020in \times 0.010in$.

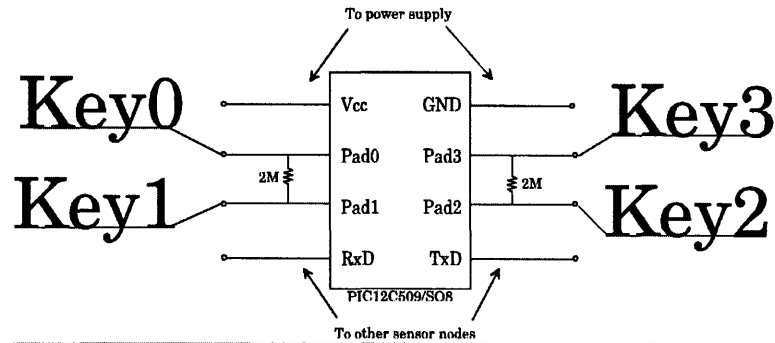


Figure B.1: Schematic of the No-Soap Radio circuit.

```

1 ;
2 ; fm455key.asm No-Soap Radio washable sewn keypad
3 ;
4 ; Copyright 1998 by Rehmi Post <rehmi@media.mit.edu>
5 ;
6 ; Created 8/18/98
7 ;
8
9     list    p=12c509,r=hex
10
11 ;
12 clockrate    equ    .4000000    ;define clock rate here
13 baudrate     equ    .1200        ;define baud rate here
14 ;
15 fclk equ      clockrate/4
16 ;*****
17 ;The value baudconst must be a 8 bit value only
18 baudconst    equ    ((fclk/baudrate)/4 - 2)
19 ;*****
20 count        equ    0x18    ;used to count tx and rx data bits
21 txreg        equ    0x19    ;used as temp transmit register
22 rcreg        equ    0x19    ;received char is saved here
23 delay        equ    0x1A    ;used to time the baud rate.

```

```
24 tempa          equ    0x1B
25
26 ;
27 ;
28 #define        _key1  GPIO,5
29 #define        _key2  GPIO,4
30 #define        _rx    GPIO,3
31 #define        _tx    GPIO,2
32 #define        _key3  GPIO,1
33 #define        _key4  GPIO,0
34 ;
35 ;
36 key1           equ    5
37 key2           equ    4
38 key3           equ    1
39 key4           equ    0
40 tx            equ    2
41 rx            equ    3
42 ;
43 KEY1          equ    020H
44 KEY2          equ    010H
45 KEY3          equ    002H
46 KEY4          equ    001H
47 TX           equ    004H
48 RX           equ    008H
49 ;
50 ;
51
52 INDF          equ    00H
53 TMRO         equ    01H
54 PCL          equ    02H
55 STATUS       equ    03H
56 FSR         equ    04H
57 OSCCAL       equ    05H
58 GPIO        equ    06H
```

```

59
60  C          equ    00H
61  DC        equ    01H
62  Z          equ    02H
63
64
65  i          equ    08H
66  j          equ    09H
67  k          equ    0AH
68  lo        equ    0DH
69  hi        equ    0EH
70  ref       equ    0FH
71
72
73  cnt        equ    10H
74  cnt2      equ    11H
75  cnt3      equ    12H
76  cnt4      equ    13H
77
78
79          ORG    0
80  reset     GOTO  main
81
82          ORG    5
83  main
84          movwf  OSCCAL
85          ; on reset the oscillator calibration value (for a nominal
86          ; 4 MHz clock) is in W. A change of +1 in OSCCAL corresponds
87          ; to approx +0.078 MHz shift in the 4 MHz clock.
88          ; So a change of +1 in OSCCAL shifts a 2-cycle square wave
89          ; generator by +9.75 kHz.
90          ; (455 kHz - 500 kHz) / (9.75 kHz) ~= -4.62
91          ; so we dither around 455 kHz by about 5 kHz by using
92          ; OSCCAL <- ref - {4,5}
93

```

```

94      movwf  ref
95      movwf  hi
96      movwf  lo
97      movlw  040H
98      subwf  hi,F          ; hi <- ref - 040H (~460 kHz)
99      movlw  050H
100     subwf  lo,F          ; lo <- ref - 050H (~450 kHz)
101
102     ;
103     ; Hi ho... Rehmi the Post here, to remind you
104     ; of one of the more despicable subtleties of
105     ; the 12C5XX, which is that GP2 will be an input
106     ; unless the TOCS bit (in OPTIONs) is cleared.
107     ;
108     movlw  08H
109     tris   GPIO
110     movlw  0DFH
111     option
112
113     goto   keyloop
114
115     loop
116     call   play_C
117     call   play_E
118     call   play_G
119     call   play_Bb
120     call   chirp
121     goto   loop
122
123
124
125     ;;
126     ;; +-----+
127     ;; | GND  GPO  GP1  GP2 |
128     ;; |  8   7   6   5  |

```

```

129      ;; |           |
130      ;; |           |
131      ;; |           |
132      ;; | 1   2   3   4 |
133      ;; \ VCC GP5 GP4 GP3 |
134      ;; \-----+
135      ;;
136
137
138      ;; to test key 1:
139      ;;      1. DRIVE pin 2 HIGH
140      ;;      2. wait 2 us
141      ;;      3. HI-Z pin 2
142      ;;      4. DRIVE pin 3 LOW then HI-Z asap (1 us)
143      ;;      5. pin 2 HIGH?
144      ;;      6. no, go to 9
145      ;;      7. (yes) increment counter
146      ;;      8. go to 4
147      ;;      9. counter > 10?
148      ;;     10. no, go to 1
149      ;;     11. (yes) play a chirp
150      ;;     12. go to 1
151      ;;
152
153
154
155      declblk macro  basename,baseaddr,num
156          local    m=0
157          while    m < num
158      basename#v(m) equ    baseaddr+m
159      m++
160          endw
161          endm
162
163

```



```
164 testkey macro key,res,num
165         local l=num
166
167         clrwdt
168         clrf cnt
169         bsf GPIO,key
170         movlw (RX|KEY1|KEY2|KEY3|KEY4)&~(1<<key)
171         tris GPIO
172
173         nop
174         movlw RX|KEY1|KEY2|KEY3|KEY4
175         tris GPIO
176
177 pulse#v(1)
178         clrwdt
179         bcf GPIO,res
180         movlw (RX|KEY1|KEY2|KEY3|KEY4)&~(1<<res)
181         tris GPIO
182         movlw RX|KEY1|KEY2|KEY3|KEY4
183         tris GPIO
184
185         btfss GPIO,key
186         goto dschgd#v(1)
187         incfsz cnt,F
188         goto pulse#v(1)
189
190 dschgd#v(1)
191         endm
192
193
194 keyloop
195         clrwdt
196         testkey key1,key2,1
197         movlw 00AH
198         subwf cnt,W
```

```
199         btfsc  STATUS,C
200         call   play_C
201
202         clrwdt
203         testkey key2,key1,2
204         movlw  00AH
205         subwf  cnt,W
206         btfsc  STATUS,C
207         call   play_E
208
209         clrwdt
210         testkey key3,key4,3
211         movlw  00AH
212         subwf  cnt,W
213         btfsc  STATUS,C
214         call   play_G
215
216         clrwdt
217         testkey key4,key3,4
218         movlw  00AH
219         subwf  cnt,W
220         btfsc  STATUS,C
221         call   chirp
222
223         goto   keyloop
224
225 chirp
226         movlw  040H
227         movwf  k
228
229 chirpl
230         movf   k,W
231         call   sing
232         decfsz k,F
233         goto   chirpl
```

```
234         return
235
236     play_C
237         movlw    010H
238         movwf    k
239     loopC
240         movlw    020H
241         call     sing
242         decfsz   k,F
243         goto     loopC
244         return
245
246     play_E
247         movlw    014H
248         movwf    k
249     loopE
250         movlw    019H
251         call     sing
252         decfsz   k,F
253         goto     loopE
254         return
255
256     play_G
257         movlw    018H
258         movwf    k
259     loopG
260         movlw    015H
261         call     sing
262         decfsz   k,F
263         goto     loopG
264         return
265
266     play_Bb
267         movlw    01DH
268         movwf    k
```

```
269  loopBb
270      movlw  011H
271      call   sing
272      decfsz k,F
273      goto   loopBb
274      return
275
276  sing
277      movwf  i
278      movwf  j
279
280      movlw  000H
281      tris   GPIO
282
283      movf   hi,W
284      movwf  OSCCAL
285  burst
286      movlw  0FFH
287  burstl1
288      xorwf  GPIO,F
289      xorwf  GPIO,F
290      xorwf  GPIO,F
291      xorwf  GPIO,F
292      xorwf  GPIO,F
293      xorwf  GPIO,F
294      xorwf  GPIO,F
295      xorwf  GPIO,F
296      xorwf  GPIO,F
297      xorwf  GPIO,F
298      xorwf  GPIO,F
299      xorwf  GPIO,F
300      xorwf  GPIO,F
301      xorwf  GPIO,F
302      xorwf  GPIO,F
303      xorwf  GPIO,F
```

```
304      xorwf  GPIO,F
305      xorwf  GPIO,F
306      xorwf  GPIO,F
307      xorwf  GPIO,F
308      xorwf  GPIO,F
309      xorwf  GPIO,F
310      xorwf  GPIO,F
311      xorwf  GPIO,F
312      xorwf  GPIO,F
313      xorwf  GPIO,F
314      xorwf  GPIO,F
315      xorwf  GPIO,F
316      nop
317      decfsz i,F
318      goto   burstl1
319
320      movf   lo,W
321      movwf  OSCCAL
322
323      movlw  0FFH
324  burstl2
325      xorwf  GPIO,F
326      xorwf  GPIO,F
327      xorwf  GPIO,F
328      xorwf  GPIO,F
329      xorwf  GPIO,F
330      xorwf  GPIO,F
331      xorwf  GPIO,F
332      xorwf  GPIO,F
333      xorwf  GPIO,F
334      xorwf  GPIO,F
335      xorwf  GPIO,F
336      xorwf  GPIO,F
337      xorwf  GPIO,F
338      xorwf  GPIO,F
```

```

339     xorwf    GPIO,F
340     xorwf    GPIO,F
341     xorwf    GPIO,F
342     xorwf    GPIO,F
343     xorwf    GPIO,F
344     xorwf    GPIO,F
345     xorwf    GPIO,F
346     xorwf    GPIO,F
347     xorwf    GPIO,F
348     xorwf    GPIO,F
349     xorwf    GPIO,F
350     xorwf    GPIO,F
351     xorwf    GPIO,F
352     xorwf    GPIO,F
353     nop
354     decfsz   j,F
355     goto     burst12
356
357     return

```

```

360 ; TABLE 3-1: PIC12C5XX PINOUT DESCRIPTION

```

```

361 ;

```

```

362

```

```

363 ; GP0          7      I/O      TTL/ST

```

```

364 ;              Bi-directional I/O port, serial programming

```

```

365 ;              data. Can be software programmed for internal

```

```

366 ;              weak pull-up and wake-up from SLEEP on pin

```

```

367 ;              change. This buffer is a Schmitt Trigger input

```

```

368 ;              when used in serial programming mode.

```

```

369

```

```

370 ; GP1          6      I/O      TTL/ST

```

```

371 ;              Bi-directional I/O port/ serial programming clock. Can

```

```

372 ;              be software programmed for internal weak pull-up and

```

```

373 ;              wake-up from SLEEP on pin change. This buffer is a

```

```

374 ; Schmitt Trigger input when used in serial programming
375 ; mode.
376
377 ; GP2/T0CKI 5 I/O ST
378 ; Bi-directional I/O port. Can be configured as T0CKI.
379
380 ; GP3/MCLR/VPP 4 I TTL
381 ; Input port/master clear (reset) input/programming
382 ; voltage input. When configured as MCLR, this pin is an
383 ; active low reset to the device. Voltage on MCLR/VPP
384 ; must not exceed VDD during normal device operation.
385 ; Can be software programmed for internal weak pull-up
386 ; and wake-up from SLEEP on pin change. Weak pull-up
387 ; always on if configured as MCLR
388
389 ; GP4/OSC2 3 I/O TTL
390 ; Bi-directional I/O port/oscillator crystal output.
391 ; Connections to crystal or resonator in crystal
392 ; oscillator mode (XT and LP modes only, GPIO
393 ; in other modes).
394
395 ; GP5/OSC1/CLKI 2 I/O TTL/ST
396 ; Bidirectional IO port/oscillator crystal input/external
397 ; clock source input (GPIO in Internal RC mode only,
398 ; OSC1 in all other oscillator modes). TTL input when
399 ; GPIO, ST input in external RC oscillator mode.
400
401 ; VDD 1 P
402 ; Positive supply for logic and I/O pins
403
404 ; VSS 8 P
405 ; Ground reference for logic and I/O pins
406
407 END

```


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